PICOSEC: Charged particle timing at sub-25 picosecond precision with a Micromegas based detector


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ABSTRACT

The prospect of pileup induced backgrounds at the High Luminosity LHC (HL-LHC) has stimulated intense interest in developing technologies for charged particle detection with accurate timing at high rates. The required accuracy follows directly from the nominal interaction distribution within a bunch crossing ($\sigma_z \sim 5$ cm, $\sigma_t \sim 170$ ps). A time resolution of the order of 20–30 ps would lead to significant reduction of these backgrounds. With this goal, we present a new detection concept called PICOSEC, which is based on a “two-stage” Micromegas detector coupled to a Cherenkov radiator and equipped with a photocathode. First results obtained with this new detector yield a time resolution of 24 ps for 150 GeV muons, and 76 ps for single photoelectrons.

1. Introduction

The prospect of pileup induced backgrounds at the High Luminosity LHC (HL-LHC) has stimulated interest in developing technology for charged particle timing at high rates [1]. Since the hermetic timing approach (where a large fraction of tracks are used to time interaction vertices) requires a large area coverage, it is natural to investigate both MicroPattern Gas and Silicon structures as candidate detector technologies to address this approach. However, since the necessary time resolution for pileup mitigation is of the order of 20–30 picoseconds (ps), both technologies require significant modification to reach the desired performance.

Photodetectors and charged particle detectors with time resolutions in the sub-nanosecond regime continue to have an impact in both High Energy physics and medical imaging. In High Energy physics the most widespread application is for particle identification, wherein the
mass of particles of known momentum is measured with the time-of-flight (TOF) particle identification technique. The current state-of-the-art technology has recently been reviewed in Ref. [2]. Existing collider experiments (e.g. ALICE) now employ large scale TOF systems with performance at the sub-100 ps level but several promising new technologies have demonstrated ~ 10 ps performance or better (for example the Microchannel PMT we use as a Cherenkov detector during the test beam measurements discussed below).

An early incentive for the development of MicroPattern Detectors (see e.g. [3]) was the promise of faster response and improved timing precision — a consequence of the more rapid signal collection and lower sensor capacitance. Except for early work at CERN (e.g. 1970–80s in the case of [3]), however, the emphasis in MicroPattern Silicon detectors rapidly moved to spatial resolution rather than temporal precision. Similarly, there has been little emphasis, in the 20 years since the GEM [4] and Micromegas [5] MicroPattern Gas Detectors (MPGD) were introduced, on exploiting their potential for fast timing. Nevertheless, in one of the original Micromegas papers [6], it was shown that sub-nanosecond time jitter could be obtained for single electrons produced at the cathode surface. This is the approach that will be followed in the current paper, i.e. the timing attributes of Micromegas are used in a photodetector.

In 2015, our collaboration proposed a structure detecting extreme ultraviolet (UV) Cherenkov light produced in a MgF	extsubscript{2} crystal coupled to semitransparent CsI photocathode and a two-stage Micromegas amplifying structure with electron amplification in both stages [7]. This detector has several advantages compared to a single-stage structure: higher gain, a reduced ion-backflow and a better separation between the electron-peak and the ion-tail of the signal. In Ref. [7], we reported encouraging results of single photoelectron timing resolution using fast laser pulses, obtained with a chamber operated in sealed mode. Subsequently, we improved the chamber integrity and the detector grounding in order to guarantee stable operation. For the results presented here we concentrate on a single pixel PICOSEC chamber (1 cm diameter), and use a bulk Micromegas amplification structure with a woven stainless steel mesh [8]. Beam tests using 150 GeV muons at CERN were carried out using several Micromegas detectors with various photocathode materials, gases, different discharge protection schemes, and read-out elements.

In this paper we present the main results obtained, demonstrating the capability of our detector to reach time resolution of tens of ps, an improvement by two orders of magnitude compared to the standard MPGD detector performance. Detailed analysis methods and simulations were developed to better understand single electron detector response and time resolution.

The manuscript is divided as follows: the PICOSEC detection concept is presented in Section 2, while the technical description of the first prototype is given in Section 2.1. The experimental setups used to measure the time resolution of single photoelectrons with the laser and 150 GeV muons in the beam are detailed in Sections 2.2 and 2.3, respectively. After a description of the individual waveform analysis in Section 2.4, the laser tests results are presented in Section 3, while those from beam tests are summarized in Section 4. The conclusions in Section 5 complete this paper.

2. The PICOSEC detection concept and the experimental setup

The detection concept presented here consists of a “two-stage” Micromegas detector coupled to a front window that acts as Cherenkov radiator coated with a photocathode, as shown in Fig. 1. A MgF	extsubscript{2} crystal is a typical radiator with light transmission down to a wavelength of 115 nm, a typical photocathode is CsI [9] with high quantum efficiency for photons below 200 nm. This configuration provides a large bandwidth for Cherenkov light production-detection in the extreme UV. The drift region is very thin (100–300 μm), which minimizes diffusion effects on the signal timing (of several ns in a typical MPGD-based drift region). Due to the high electric field, photoelectrons also undergo pre-amplification in the drift region.

The readout is a bulk Micromegas [8], which consists of a woven mesh (18 μm-diameter wires) and an anode plane, separated by a gap of 128 μm, mechanically defined by pillars. This type of readout, operated in Neon- or CF	extsubscript{4}-based gas mixtures, can reach gains of 10	extsuperscript{5}–10	extsuperscript{6}, high enough to detect single photoelectrons [10]. Moreover, the electron drift velocity is high enough for a fast charge collection: ranging from 9 to 16 cm/μs for drift fields between 10 and 20 kV/cm in the “COMPASS” gas (80%Ne + 10%C	extsubscript{2}H	extsubscript{6} + 10%CF	extsubscript{4}), according to simulations with Magboltz [11].

In normal operation, a relativistic charged particle traversing the radiator produces UV photons, which are simultaneously (RMS less than 10 ps) converted into primary (photo) electrons at the photocathode. These primary electrons are preamplified in the drift region due to the high electric field (~20 kV/cm); then, they partially traverse the mesh (~25% due to the field configuration), and are finally amplified in the amplification gap, where a high electric field (~40 kV/cm) is applied.

The arrival of the amplified electrons at the anode produces a fast signal (with a risetime of ~0.5 ns) referred to as the electron-peak, while the movement of the ions produced in the amplification gap generates a slower component-ion-tail (~100 ns). A typical waveform is shown in Fig. 2. The maximum drift time of ions is below 630 ns, which is low enough not to affect the detector rate capability.

It should be noted that due to preamplification in the thin drift gap the relative contribution to the overall signal of direct ionization produced by the traversing particle is negligible. In the COMPASS gas and for the conditions described in Section 2.3, relativistic muons create ~ 21 ion clusters/cm with few ionization electrons per cluster. The probability to produce enough ionization charge that undergoes the same amplification (i.e. in the first ~30 μm) as the typical 10 photoelectrons from the Cherenkov signal is only a few percent.
A 1 mm thick quartz crystal with a deposit of 10 nm Aluminum, or 100 nm MgF₂ layer was deposited over the chromium substrate in order to increase transparency. At the bottom part of the PCB, there is a 18 μm copper layer used as ground reference. The amplification gap (between the mesh and the anode in Fig. 1) is defined by only six 200 μm-diameter wires and has a 51% optical transparency. Four kapton rings, of 50 μm thickness each, are placed between the mesh and the crystal (i.e. radiator) to define the height of the drift gap of 200 μm. During the laser tests, the crystal was made out of a 3 mm thick MgF₂, on which a 5.5 nm chromium layer was deposited serving as photocathode. During beam tests, an additional 18 nm-thick CsI layer was deposited over the chromium substrate in order to increase the number of photoelectrons produced by charged particles. In both configurations, a 10 nm-thick metallic ring is placed on the crystal to establish the potential of the photocathode.

The whole detector is installed inside a stainless steel chamber, which is then filled with the COMPASS gas at 1 bar absolute pressure. Other gases, like 80%CF₂ + 20%C₄H₆ at 0.5 bar absolute pressure, have also been used but this article will focus on the results obtained with the COMPASS gas. The vessel has a transparent (quartz) entrance window to allow the passage of either UV light or laser pulses; it has two gas valves for gas circulation, as well as a large vacuum port for evacuating the vessel.

Referring to Fig. 4, cathode, anode and mesh elements are electrically connected by SMA or SHV feedthroughs, as indicated. The cathode is connected to one CAEN High Voltage Supply (HVS) channel, the anode to a CIVIDEC preamplifier (2 GHz, 40 dB, a gain of 100), biased by a 500 m long BNC cable, terminated with a 50 Ohm resistor in order to avoid signal reflections. Special attention was paid to proper grounding throughout the electronics design, and referred to the ground layer of the Micromegas readout. Each of the high voltage lines has a dedicated low-pass filter to suppress ripples from the HVS.

The time response of the PICOSEC detector for single photoelectrons was measured at the Saclay Laser-matter Interaction Center (IRAMIS/SLIC, CEA). The experimental setup (Fig. 5) includes a femtosecond laser with a pulse rate ranging from 9 kHz to 4.7 MHz at 267–288 nm wavelength and a focal length of ~1 mm. The laser beam is split into two equal parts, one arriving directly at the prototype and the other at a fast photodiode (PD0). The PD0 signal serves as the reference time, simultaneously recorded with the PICOSEC detector signal during the measurements. The high laser intensity at the PD0 and the fast risetime of this device result in a reference time accuracy of approximately 13 ps. The intensity of the laser arriving at the detector is reduced by a series of light attenuators: electroformed fine nickel meshes (100–2000 LPI) with optical transmission varying between 10% and 25%, yielding attenuation factors of 4, 5, 10, and their combinations. The Micromegas detector signal goes through a CIVIDEC preamplifier before being digitized and registered together with the PD0 signal by a 2.5 GHz oscilloscope at a rate of 20 GSamples/s (i.e. one sample every 50 ps). The PICOSEC detector was operated with the COMPASS gas at 1 bar absolute pressure. The anode voltage (HV2 in Fig. 1) was scanned between 450 V and 525 V in steps of 25 V, while the drift voltage (HV1) was varied in steps of 25 V in different ranges, depending on the anode voltage. These experimental conditions, voltages used and the measured time resolution, are summarized in Table 1. In general, the lowest voltage used corresponds to a detector gain high enough to distinguish the signal from the noise level (gain ~ 10³), while the highest voltage is the maximum value for which the detector operates in stable conditions (up to gains of ∼ 10⁷). For each voltage configuration, more than 10⁸ events were recorded with the oscilloscope, and subsequently analyzed offline.

During the first data-taking campaign, the photocathode efficiency was less than 0.5%, which led us to derive the trigger directly from the PICOSEC signal, in the interest of data collection efficiency. Results are based on this first campaign as runs contain enough events for the analysis. The PICOSEC trigger threshold varied between 10 and 90 mV, as shown in Table 1, leading to a bias of the recorded PICOSEC pulse height spectrum. In a later data-taking campaign, the photocathode...
was replaced to increase the signal efficiency up to \( \sim 5\% \); this allowed deriving the trigger decision from the fast photodiode. The detector was operated in the same voltage conditions as the dataset with a trigger bias on the PICOSEC amplitude, so they could be used to confirm that the charge distribution follows a Polya function \([12]\), as discussed in Section 3.

2.3. Beam tests with 150 GeV muons

The time response of the detector to 150 GeV muons was measured during several beam periods in 2016 and 2017 at the CERN SPS H4 secondary beamline. The experimental setup (Fig. 6) allows the characterization of up to three PICOSEC detectors, situated at the positions Pos0, Pos1 and Pos2. Two trigger scintillators of \( 5 \times 5 \) mm\(^2\) operate in anti-coincidence with a veto scintillator whose aperture (hole) matches the same area. This trigger configuration efficiently selects muons that do not undergo scattering and suppresses triggers from particle showers. One Hamamatsu MCP PMT\(^6\) is used as time reference; its entrance window (3 mm thick quartz) is placed perpendicular to the beam and serves as Cherenkov radiator. From the time jitter between two identical MCPs we determine the reference time accuracy of \( 5.4 \pm 0.2 \) ps. A telescope of three tracking GEM detectors with two-dimensional strip readout is used to reconstruct the trajectory of each muon with a combinatorial Kalman filter based algorithm, and to determine its impact position at each detector.

The PICOSEC (and MCP reference) waveforms are recorded in the beam tests in the same way as in the laser tests using an oscilloscope with 20 GSamples/s and 2.5 GHz bandwidth, while the tracking (GEM detector) data are recorded simultaneously in an APV25 based SRS DAQ \([13]\). To ensure event alignment in the two DAQ systems, the internal SRS event number is sent as a bit stream to one oscilloscope channel. The DAQ trigger is generated by the scintillators.

Each PICOSEC detector is operated at different anode and drift voltages, which are respectively scanned in steps of 25 V. As in the case of laser tests, the minimum and maximum drift voltages respectively correspond to the cases where the signal is distinguishable from the noise level and the detector operates in stable conditions. However, the detector gain \( \left( 10^{4} - 10^{5} \right) \) was lower than in laser tests because the initial number of electrons was higher. For a fixed anode voltage, the drift voltage was \( \sim 100 \) V lower than in laser tests. For each voltage configuration, more than 4000 events are recorded with the oscilloscope, and subsequently analyzed offline. During periods without beam or long accesses, each detector is illuminated with a UV lamp to measure the response to single photoelectrons at different voltage settings. This information is later analyzed to estimate the mean number of photoelectrons produced by muons during beam tests.

2.4. Waveform analysis

In this section, we briefly describe the analysis performed on both laser and beam test data. For each PICOSEC signal, the baseline offset and noise level are determined using the 75 ns precursor of the pulse. Then, the “electron-peak” amplitude \( V_{\text{max}} \) is defined as the difference between the highest point of the waveform and the baseline. For the timing measurement, a Constant Fraction (CF) method based on a sigmoid function is used to minimize the contribution of the noise. Other algorithms to determine the CF have been used with similar results.

In this approach, a sigmoid function is fit to the leading edge of the electron-peak. This function is defined as

\[ V(t) = \frac{P_0}{1 + \exp \left( -P_1 \times (t - P_2) \right)} + P_3 \]  

where \( P_0 \) and \( P_3 \) are respectively the maximum and the minimum values, \( P_1 \) is the inflection time (i.e. where the slope changes derivative), and \( P_2 \) quantifies the speed of the sigmoid change (i.e. is correlated to the signal risetime). The time corresponding to a 20% CF is calculated as follows:

\[ t_2 = P_1 - \frac{1}{P_2} \log \left( \frac{P_0}{0.2 \times V_{\text{max}} - P_3} - 1 \right) \]  

For the photodetectors used as the time reference (i.e. MCP, or PD0 in the case of laser tests) a simpler approach is applied as signals are almost immune to any source of noise: after the calculation of the pulse baseline and amplitude, a cubic interpolation between four points around CF=20% is used to extract the temporal position of the signal with better precision. The “Signal Arrival Time” (SAT) is then defined as the difference between the PICOSEC CF time and that of the reference detector, as illustrated in Fig. 2.

The “electron-peak charge” is defined as the integral of the waveform between the start and the end points, defined as the first points situated before and after the maximum whose amplitude is less than one standard deviation away from the baseline offset. For those pulses with no clear separation between the electron-peak and the ion-tail, the end point has been alternatively defined as the time when the pulse derivative changes sign. The resulting value is then transformed to Coulombs, using the measured electron-peak charge-to-amplitude ratio of 0.833 pC/mV.

3. Laser test results

Two aspects of the PICOSEC time response in the laser measurements are discussed below. Firstly, we discuss the dependence of the time response on signal amplitude as this dependence (particularly concerning the role of the drift field and fluctuations in the preamplification at a given field) elucidates the physical origin of the PICOSEC time resolution. Secondly, we convolute this amplitude dependence with the observed scaling with the pulse amplitude for fixed anode voltage (c.f. Fig. 7 (left)), that reflects the dependence of the electron drift velocity on drift field, and that b) the time resolution properties of the PICOSEC detector are described by a single function and are mainly determined by the electron-peak charge.

In fact there are two physical parameters of the drift region which could, through their dependence on the applied drift voltage, affect the timing performance: a) the longitudinal diffusion coefficient (i.e. \( \sigma_l \sim E_{\text{field}}^{1/2} / v_{\text{drift}} \)) and b) the mean free path to the first ionizing collision. The observed scaling with the pulse amplitude for fixed anode voltage suggests that the improvement of the time resolution is driven by the latter effect (i.e. (b)), while the contribution from the variation in longitudinal diffusion is less significant in this regime.

Yet another indication of the dominant role of the drift region can be derived from Fig. 8, where the dependence of the time resolution on electron-peak charge is shown for different anode voltages. To
preserve the same level of signal amplitudes at lower anode voltages, the drift fields have been correspondingly increased. Signals with the same electron-peak charge and lower anode voltage (and thus necessitating a higher preamplification) show a better time resolution, i.e., pulses with a higher preamplification gain have better timing properties than those with a higher amplification gain.

The CF algorithm discussed above is used to eliminate the expected correlation between signal amplitude and SAT observed for signals with similar shapes but different amplitudes (known as “time walk correction” [14]) normally observed when timing is derived from a fixed threshold. Nevertheless, there are also well known examples where both amplitude and signal risetime can vary from pulse to pulse, requiring “amplitude and risetime correction” for SAT determination. We also considered this hypothesis since the time resolution varies by several hundreds of picoseconds, for different signal amplitudes — even with the CF method. However, as shown in Fig. 9, the average electron-peak shape remains essentially identical for different electron-peak charges. For this reason, the correlation between electron-peak charge and the signal arrival time observed in Fig. 9 must be a consequence of the physical mechanism generating the PICOSEC signal rather than an artifact of the timing algorithm.

3.1. Derivation of the overall “single photoelectron time distribution function”

As described in Section 2.2, data are collected with an electronic trigger generated by the PICOSEC detector for part of the dataset. The threshold level was in some cases high in comparison to the Root Mean Square (RMS) baseline noise (typically ∼2.5 mV), as detailed in Table 1. Supposing that the derived dependence of the mean and
standard deviation of the SAT with the electron-peak charge are also valid for pulse amplitudes lower than the threshold, the time resolution of the PICOSEC detector signal at a given operating point is estimated by the equation:

\[ \sigma^2 = \sum_{i=1}^{N} a_i \sigma_i^2 + \sum_{j=1}^{n} a_j x a_j \left( \sigma_j^2 + \sigma_e^2 + (\mu_j - \mu_e)^2 \right) \]  

(4)

where \( \mu_i \) and \( \sigma_i \) are the mean and standard deviation of the SAT in an interval \( i (i = 1, \ldots, n) \) of the electron-peak charge \( Q_i \), and \( a_j \) is the probability density function (PDF) for a given charge \( Q_j \), i.e. \( a_j = A(Q_j) \), and \( \sum_{i=1}^{N} a_i = 1 \). The term \( (\mu_j - \mu_e) \) takes into account the difference in SAT at different electron-peak charges, caused by the measured correlation between these two variables. For all cases, the minimum electron-peak charge \( Q_e \) is set to 0.33 pC (i.e. 10 mV in amplitude), equivalent to four times the typical RMS baseline noise. Meanwhile, the \( A(Q) \) PDFs are obtained by fitting each electron-peak charge distribution by a Polya function [12] which is expressed as:

\[ A(Q|N, Q_e, \theta) = \frac{(\theta + 1)^{N(\theta+1)}}{(N(\theta+1))} \left( \frac{Q}{Q_e} \right)^{N(\theta+1)-1} \exp \left[ -(\theta + 1) \frac{Q}{Q_e} \right] \]  

(5)

where \( Q_e \) is the mean charge per single photoelectron, \( N \) is the number of photoelectrons, and \( \theta \) is the Polya shape parameter. This function describes well the single electron-peak charge response of the PICOSEC detector \( (N = 1) \), as shown in Fig. 10, including also the dataset without a PICOSEC trigger threshold bias (Fig. 10 right). In each fit, bin sizes and fitting regions were varied in order to estimate the systematic errors, which were then combined with the statistical uncertainties.

The raw and corrected time resolution values of the PICOSEC detector for single photoelectron detection at the different operation settings are shown in Table 1. Two uncertainties are included in the calculation: the uncertainty in the parameterization of the mean and of the standard deviation of the SAT with the charge, and the uncertainty in the Polya parameterization. The best measured value of the time resolution is 76.0 ± 0.4 ps, which is obtained for the lowest applied anode voltage (450 V). Meanwhile, as can be seen from the dependence of the time resolution on the drift and anode voltages (Fig. 11), the time resolution is better for higher drift voltage. We did not explore the whole parameter space but a further improvement is expected if the drift voltage can be further increased, while keeping the gain almost constant (by reducing the anode voltage correspondingly). In fact, a simulation of the detector response [15] has shown that the detector time jitter is mainly defined by the drift (pre-amplification) stage, while the contribution of the amplification stage is negligible.

### 4. Beam tests results with 150 GeV muons

The same analysis as in Section 3 is applied to the SAT distributions of 150 GeV muons, as a correlation with the electron-peak charge is expected. However, as shown in Fig. 12 (left), the mean of the SAT distribution is almost constant for each setting; this is explainable by the high drift fields (and preamplification gains) at which the PICOSEC detector is operated. Meanwhile, the time resolution decreases as the electron-peak charge increases (Fig. 12, right).

As the mean of the SAT distribution is almost independent of the electron-peak charge, each SAT distribution generated by 150 GeV muons is fit by a two Gaussian distribution (both Gaussians centered at the same value) and the time resolution is reported as the standard deviation of the full distribution. The time resolution results obtained

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7 A single Gaussian fit has also been used with similar results.
The mean number of photoelectrons \(\nu\) is the number of values) were then used to define a likelihood function

\[
\mathcal{L}(N|\bar{Q}_i, \theta) = \prod_{i=1}^{n} \frac{N!}{\bar{Q}_i! \times A(Q_i|\bar{Q}_i, \theta)}
\]

(6)

Table 1
Laser tests: The experimental conditions, voltages used and the time resolution (raw and corrected values) of the PICOSER detector, operated with the COMPASS gas at 1 bar absolute pressure. The threshold is the amplitude of the smallest signal recorded at the trigger level and the RMS is the standard deviation of the baseline. The raw time resolution values are estimated using the CF algorithm, while the corrected values are estimated after correcting the correlation of the SAT with the electron-peak charge, as discussed in detail in Section 3.1.

<table>
<thead>
<tr>
<th>Anode (V)</th>
<th>Drift (V)</th>
<th>Threshold (mV)</th>
<th>RMS (mV)</th>
<th>Time resolution (ps)</th>
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</table>

are as low as 24.0 ± 0.3 ps, as shown in Fig. 13 for anode and drift voltages of 275 V and 475 V, respectively.

From a scan over a wide range of voltage settings we obtain the dependence of the time resolution on the drift and anode voltages, as shown in Fig. 14. This figure clearly shows that the time resolution improves for higher drift voltages, while the gain is kept constant by reducing in the same proportion the anode voltage. The optimal time resolution is reached for drift voltages of 450–475 V, which are the maximum settings at which the detector can be stably operated, i.e. there is no discharge during the beam run.

The mean number of photoelectrons \(N\) is estimated for those voltage settings, from the response to a single photoelectron calibration using the UV lamp. In a first step, the electron-peak charge distribution of the UV lamp runs is fit by Eq. (5) (where \(N = 1\), in order to estimate the parameters \(Q\) and \(\theta\). The electron-peak charge values \((Q_i)_{i=1}^n\) of the 150 GeV muon run (where \(n\) is the number of values) were then used to define a likelihood function

\[
\mathcal{L}(N|\bar{Q}_i, \theta) = \prod_{i=1}^{n} \frac{N!}{\bar{Q}_i! \times A(Q_i|\bar{Q}_i, \theta)}
\]

(6)
where $Q_e$ and $\theta$ are results of previous fit, $N$ is the mean number of photoelectrons per muon including the geometrical acceptance and $A(Q)$ is the Poly-a-function defined in Eq. (5). This function was then maximized in order to estimate $N$. An example of the results of these two fits is shown in Fig. 15, while the value obtained is $N \sim 10.4 \pm 0.4$. The uncertainty of this estimation is dominated by the fit uncertainty of the UV lamp runs.

5. Conclusions

In this paper, we present a new detector concept, called PICOSEC, composed of a “two-stage” Micromegas detector coupled to a Cherenkov radiator and equipped with a photocathode. The good timing resolution performance for single photoelectrons ($\sigma_t \sim 76$ ps) and for 150 GeV muons ($\sigma_t \sim 24$ ps) is promising and motivates further development towards practical applications. Among the significant issues to be addressed to ensure suitability for a large area detector to be used in high rate experiment are: 1) the development of efficient and robust photocathodes (or secondary emitters), and 2) scalability, including the development of the corresponding readout electronics.

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